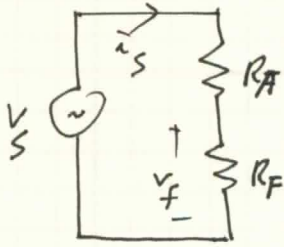


* 2-resistor feedback (series) ~~series~~

equivalent
feedback
system

i) Without R_F , i_S will be developed as

$$i_S = \frac{V_S}{R_A}$$

ii) When R_F included,

R_F will detect i_S and develop V_F .

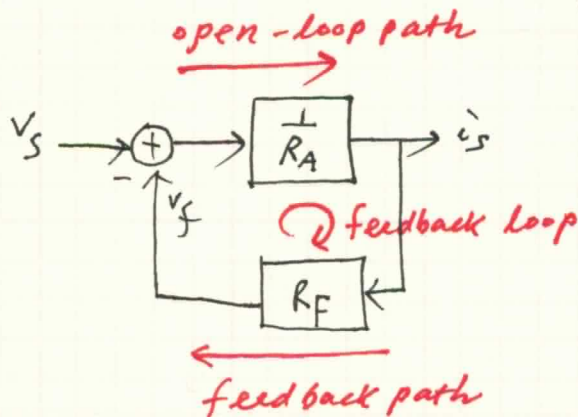
And the V_F will be subtracted from V_S .

$$\text{The resultant } i_S = \frac{V_S - V_F}{R_A} \quad \text{--- (A)}$$

$$\Rightarrow V_F = i_S R_F \quad \text{--- apply to (A)}$$

$$\Rightarrow i_S R_A = V_S - V_F = V_S - i_S R_F$$

$$\therefore i_S = \frac{V_S}{R_A + R_F} = \frac{V_S}{R_A} \frac{1}{\left(1 + \frac{R_F}{R_A}\right)}$$



open loop gain, $A_{op} = \frac{1}{R_A}$
feedback factor, $f = R_F$
loop gain, $T = A_{op} f$
 $= \frac{R_F}{R_A}$

closed-loop gain (A_f)

$$= \frac{i_S}{V_S} = \frac{A_{op}}{1 + T} = \frac{A_{op}}{1 + A_{op} f}$$

$$= \frac{\frac{1}{R_A}}{1 + \frac{R_F}{R_A}} = \frac{1}{R_A + R_F}$$

⇒ Without R_F , $Z_{in} = R_A$

With R_F , $Z_{in} = R_A + R_F = R_A \left(1 + \frac{R_F}{R_A}\right)$
 $= R_A (1 + T)$

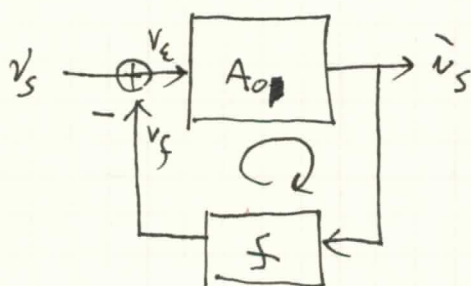
Z_{in} increases
by a factor of
(1+T).
(why?)

⇒ Without R_F , $Z_{out} = R_A$

With R_F , $Z_{out} = R_A + R_F = R_A \left(1 + \frac{R_F}{R_A}\right)$
 $= R_A (1 + T)$

Z_{out} increases
by a factor of
(1+T).
(why?)

< Generalization >



$$v_s - v_f = v_e$$

$$= \frac{1}{1+T} v_s$$

NOTE: input is Voltage
output is Current

① Loop gain $= T = A_{op} \cdot f$

② closed loop gain $= A_f = \frac{A_{op}}{1+T} = \frac{A_{op}}{1+A_{op}f}$

1st-order approximation \leftarrow

$$\approx \frac{1}{f} \left(1 - \frac{1}{T}\right) \quad (\text{if } T \gg 1) \quad = \frac{1}{f} \left(\frac{1}{1 + \frac{1}{A_{op}f}}\right) = \frac{1}{f} \left(\frac{1}{1 + \frac{1}{T}}\right)$$

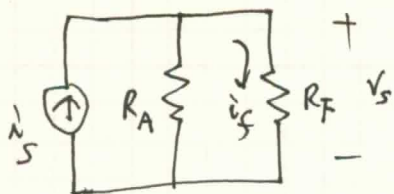
2nd-order approximation \leftarrow

$$\approx \frac{1}{f} \quad (\text{if } T \gg 1)$$

③ $Z_{in,f} = Z_{in} (1+T)$ \leftarrow makes more ideal voltage driving.

④ $Z_{out} = Z_{out} (1+T)$

⊛ 2-resistor feedback (parallel)



equivalent
feedback
system.

i) Without R_F , V_s will be developed as

$$V_s = i_s R_A.$$

ii) when R_F paralleled,

R_F will detect V_s and generate i_f .

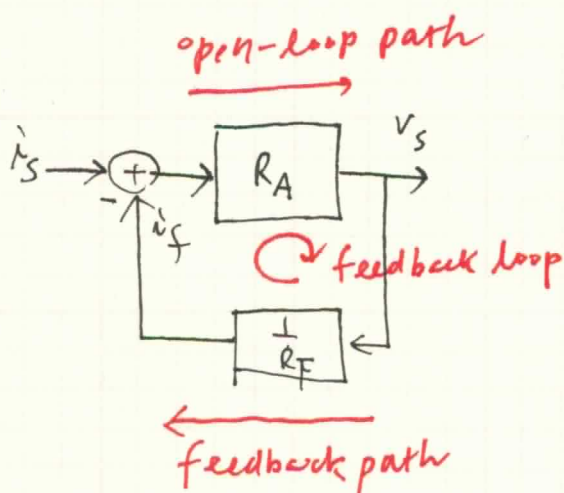
And the i_f will be subtracted from i_s .

The resultant $V_s = (i_s - i_f) R_A$ — (A)

$$\Rightarrow i_f = \frac{V_s}{R_F} \quad \text{apply to (A)}$$

$$\Rightarrow V_s = \left(i_s - \frac{V_s}{R_F} \right) R_A$$

$$\therefore V_s = \frac{i_s R_A}{1 + \frac{R_A}{R_F}}$$



open loop gain, $A_o = R_A$
feedback factor, $f = \frac{1}{R_F}$

$$\text{loop gain, } T = A_o f = R_A / R_F$$

Closed loop gain (A_f)

$$= \frac{V_s}{i_s} = \frac{A_o}{1 + T} = \frac{R_A}{1 + \frac{R_A}{R_F}}$$

$$= \frac{R_A \cdot R_F}{R_A + R_F} = R_A \parallel R_F$$

\Rightarrow without R_F , $Z_{in} = R_A$

with R_F , $Z_{in} = R_A \parallel R_F = \frac{R_A \cdot R_F}{R_A + R_F} = \frac{R_A}{1 + \frac{R_A}{R_F}}$

Z_{in} decreases
by a factor of
(1+T).

(Why?)

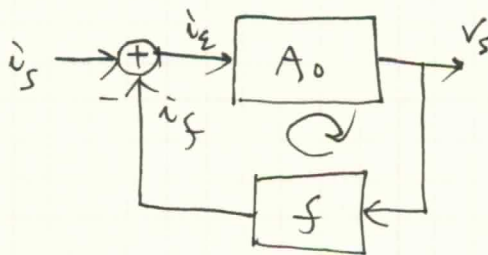
$$= \frac{R_A}{1+T}$$

\Rightarrow without R_F , $Z_{out} = R_A$

with R_F , $Z_{out} = R_A \parallel R_F = \frac{R_A}{1 + \frac{R_A}{R_F}} = \frac{R_A}{1+T}$

Z_{in} decreases
by a factor of (1+T).
(Why?)

< generalization >



$$\dot{v}_s - \dot{v}_f = \dot{v}_e = \frac{1}{1+T} \times \dot{v}_s$$

NOTE: Input is current
output is voltage

① loop gain, $T = A_o f$

② closed loop gain, $A_f = \frac{A_o}{1+T} = \frac{A_o}{1+A_o f}$

$$= \frac{1}{f} \left(\frac{1}{1 + \frac{1}{A_o f}} \right) = \frac{1}{f} \left(\frac{1}{1 + \frac{1}{T}} \right)$$

$$\approx \frac{1}{f} \left(1 - \frac{1}{T} \right) \quad \text{③ } T \gg 1$$

$$\approx \frac{1}{f}$$

③ $Z_{in f} = Z_{in} \frac{1}{1+T}$

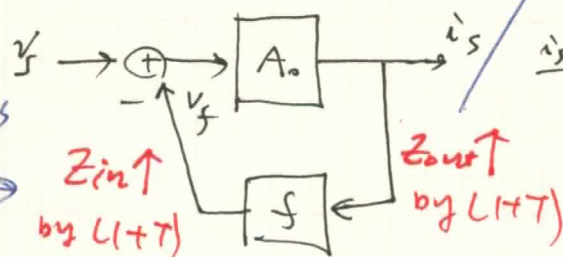
← makes more ideal
current driving.

④ $Z_{out f} = Z_{out} \frac{1}{1+T}$

Feedback Basics

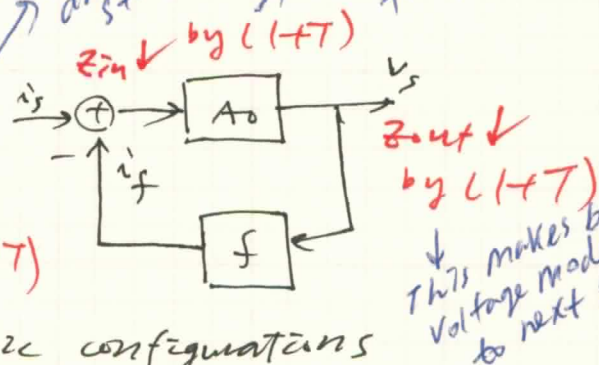
(*) feedback configurations

this makes better voltage driving for previous stage.

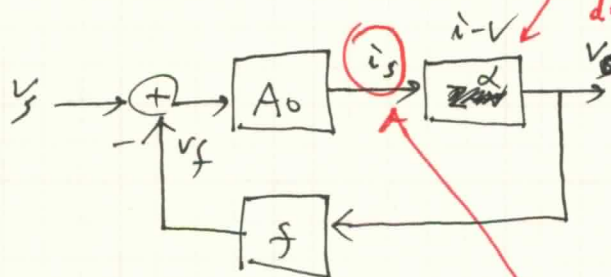


this makes better current driving to next stage.

this makes better current mode driving for previous stage.



- \Rightarrow these are two basic configurations
- \Rightarrow driving source should be either voltage or current.
- \Rightarrow if driving source is voltage, immediate output response is current, which is first case.
- \Rightarrow if driving source is current, immediate output response is voltage, which is the second case.

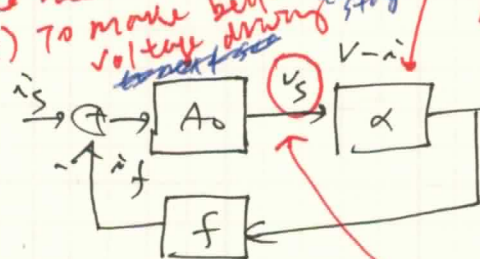


- \Rightarrow $i-v$ converter can be used to convert i_s to v_o .
- \Rightarrow But, still the point of feedback operation is to decrease i_s .
- \Rightarrow decreasing i_s results in decreasing v_o .

$$\Rightarrow Z_{in} \uparrow \times (1+T)$$

$$Z_{out} \downarrow \times (1+T)$$

why do we need this? A) To make better voltage driving to next stage.



why do we need this? A) To make better current driving to next stage.

- \Rightarrow $v-i$ converter can be used to convert v_s to i_o .
- \Rightarrow But, still the point of feedback operation is to decrease v_s .
- \Rightarrow decreasing v_s results in decreasing i_o .

$$\Rightarrow Z_{in} \downarrow \times (1+T)$$

$$Z_{out} \uparrow \times (1+T)$$